

ACOUSTIC DATA PROCESSING USING THE DECENTRALIZED SQUARE ROOT INFORMATION FILTER

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ACOUSTIC DATA PROCESSING USING THE DECENTRALIZED SQUARE ROOT INFORMATION FILTER

Project Summary

Very Large Scale Integration (VLSI) technology has been developed to the point where special purpose processors may be concatenated to form supercomputers with far greater throughput rates than uniprocessor machines. MTI has developed a parallel form of the conventional Kalman filter that is well suited to being implemented in a multiprocessing environment. Moreover, our Decentralized Square Root Information Filter (DSRIF) has several very unique features which could be incorporated into the design of an integrated undersea tracking system with much improved performance over existing methods.

Phase I research demonstrated feasibility of the DSRIF as a means for solving the linear least squares estimation problem in decentralized form. Underwater tracking of a high velocity torpedo (undergoing high dynamic maneuvers) was simulated. Also, an extended form of the DSRIF was used to complete the processing of real Multiple Rocket Launch System Data provided by the White Sands Missile Range. In this case, an adaptive form of the extended DSRIF, wherein the process noise levels were made to be a function of the globally optimal estimate error covariance, was successfully used to track the rocket data.

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LIST OF SYMBOLS

- i superscripted local system number
- j superscripted vector element number
- k time index, may be subscripted or enclosed in parentheses
- x_k global state vector
- $w_{\mathbf{k}}$ global process noise vector
- $\Phi_{\mathbf{k}}$ global state transition matrix
- x_k^i local state vector
- oxⁱ origin of a local coordinate system in earth centered earth fixed coordinates
- ox origin of the global coordinate system in earth centered earth fixed coordinates
- orⁱ radial vector from earth's center to the origin of a local coordinate system
- $w_k^{\underline{i}}$ local process noise vector
- Φ_{ν}^{i} local state transition matrix
- y_k^i local measurement vector
- Ck global observation sub-matrix
- $H_{\mathbf{k}}^{\mathbf{i}}$ local observation matrix
- v_{k}^{i} local measurement noise vector
- $x_k(\underline{+})$ (measurement updated) global state estimate time updated
- $x_{k}^{i}(\pm)$ (measurement updated) local state estimate time updated
- $P_{k}(\pm)$ (measurement updated) global estimate error covariance matrix time updated
- $P_{k}^{i}(\underline{+})$ (measurement updated) local estimate error covariance matrix time updated
- $\mathbf{M}_{\mathbf{K}}^{\mathbf{i}}$ matrix of zeros and ones which partition the global states to the local systems
- $Q_{\mathbf{k}}$ covariance matrix for process noise vector

- $R_{\mathbf{k}}^{\mathbf{i}}$ covariance matrix for local measurement noise vector
- $P_0(-)$ initial global estimate error covariance matrix
- N number of time samples
- M number of local systems
- Mⁱ number of local measurement variables
- $z_w(k)$ "pseudomeasurement" vector
- $R_{\mathbf{w}}(\mathbf{k})$ inverse square root of $Q_{\mathbf{k}}$
- $R_{\mathbf{k}}(\underline{+})$ (measurement updated) global square root information matrix time updated
- $z_k(\underline{+})$ (measurement updated) global square root information vector time updated
- $R_{K}^{i}(\underline{+})$ (measurement updated) local square root information matrix time updated
- $z_{k}^{i}(\underline{+})$ (measurement updated) local square root information vector time updated
- e_k^i local "innovations" vector
- (i) (i) (i) $R_{w}^{*}(k) R_{wx}^{*}(k) z_{w}^{*}(k)$ local smoothing coefficients
- $R_{\mathbf{w}}^{*}(\mathbf{k})$ $R_{\mathbf{w}\mathbf{x}}^{*}(\mathbf{k})$ $z_{\mathbf{w}}^{*}(\mathbf{k})$ global smoothing coefficients
- $z^*(k) H^*(k)$ merge coefficients
- t continuous time variable
- dⁱ local coordinate system displacement vector
- L latitude of global coordinate system origin
- 8 longitude of global coordinate system origin
- α^1 β^i Euler angles defining the orientation of the local coordinate system τ^i w.r.t. to the global coordinate system
- re average earth radius
- ${}^{1}\mathsf{T}_{k}^{i}$ orthogonal transformation in local time update
- $^2\mathsf{T}^{\mathbf{i}}_{\mathbf{k}}$ orthogonal transformation in local measurement update

- ${}^{\bf 3}{\bf T}_{\bf k}$ —orthogonal transformation in first merge step
- $^4\text{T}_{\mathbf{k}}$ $\,$ orthogonal transformation in second merge step
- $\mathsf{T}^{\dot{\mathsf{1}}}$ orthogonal coordinate transformation between the local and global coordinate system
- orthogonal coordinate transformation between the global coordinate system and the earth centered earth fixed coordinate system
- μ product of G and re
- r_k^i range measurement
- $\dot{r}_{\mathbf{k}}^{\mathbf{i}}$ range rate measurement
- θ_k^i elevation angle measurement
- $\Gamma_{\mathbf{k}}^{\mathbf{i}}$ azimuth angle measurement
- G gravitational constant
- m_e earth's mass
- Ω earth's angular velocity
- u(t) deterministic control
- f(x(t),u(t)) nonlinear dynamics vector
- $h^{i}(x(t))$ nonlinear local observation vector
- $F_{\mathbf{k}}$ linearized dynamics matrix

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1.0 Introduction and Problem Statement

The trajectory estimation problem is a problem of nonlinear estimation. Assume that the state of a target evolves in time according to the equation

$$\dot{x}(t) = f(x(t), u(t)) + w(t) \tag{1}$$

where u(t) is the target's nominal control vector and w(t) is a zero mean white noise process with spectral density Q(t). Values for the latter are selected in order to compensate for errors in the model which may originate from unknown perturbations to the nominal control (such as turbulence, strong currents, large surface waves, etc.) as well as from uncertainties in the hydrodynamics of the target vehicle. The corresponding discrete measurement vector for the i^{th} sensor is given by

$$y_k^i = h^i(x_k) + v_k^i \tag{2}$$

where v_k^i is a zero mean white noise sequence with covariance R_k^i . The problem is to estimate the target states x_k based upon all of the past measurements y_1^i where $\{1 \le i \le M \text{ and } 1 \le 1 \le k\}$. The state vector x_k contains the target position and velocity, biases which account for the displacement and orientation of the sensor or "local" coordinate systems (LCSs) w.r.t. the global coordinate system (GCS), and acceleration when the target vehicle is maneuvering.

Unfortunately, the optimal nonlinear estimator (conditional mean) cannot be realized with a finite-dimensional implementation and consequently, all practical nonlinear filters must be suboptimal. The usual suboptimal solution is the Conventional Kalman Filter when the nominal trajectory is known a priori, the Extended Kalman Filter when the nominal trajectory is unavailable, and Higher Order Filters (such as the second order filter [3], the single-stage iterative filter [4], and the Gaussian sum filter [5], among others [6]) when even greater accuracy is desired. The tradeoff here is performance versus real-time computational requirement.

Thus, we see that the ability of a sensor group to accurately record the motion of one or more underwater targets is a function of the individual target and sensor dynamical models as well as the particular algorithm used to combine raw data and produce track estimates. Another issue is intersensor communication. Sensors which operate independently from one another will exhibit larger estimate errors than ones which communicate with other members of the network.

Let the global discrete time linear system

$$x_{k} = \Phi_{k-1} x_{k-1} + w_{k-1}$$
 (3)

where

$$w_{k} = N(0, Q_{k}) \tag{4}$$

$$x_0 = N(0, P_0(-))$$
 (5)

be the target dynamics linearized about the nominal trajectory (or estimate), and the global measurement model

$$\begin{bmatrix} v_k^1 \\ \vdots \\ v_k^M \end{bmatrix} = \begin{bmatrix} c_k^1 \\ \vdots \\ c_k^M \end{bmatrix} \times_k + \begin{bmatrix} v_k^1 \\ \vdots \\ v_k^M \end{bmatrix}$$

$$\begin{bmatrix} v_k^1 \\ \vdots \\ v_k$$

where v_k^1 through v_k^M are uncorrelated random vectors and

$$v_k^i = N(0, R_k^i) \tag{7}$$

be the model for the tracking sensors.* The problem is to calculate the globally optimal (minimum mean square error) estimate of \mathbf{x}_k and its associated estimate error covariance matrix \mathbf{P}_k when \mathbf{y}_k^1 through \mathbf{y}_k^M are processed separately by locally optimal estimators 1 through M correspondingly. The local dynamical models are

$$x_{k}^{i} = \Phi_{k-1}^{i} x_{k-1}^{i} + w_{k-1}^{i}$$
 (8)

where

$$w_k^i = N(0, Q_k^i) \tag{9}$$

$$x_0^i = N(0, P_0^i(-))$$
 (10)

and the local measurement models are

$$y_k^i = H_k^i x_k^i + v_k^i \tag{11}$$

where v_k^i satisfies (7). Notice that the local states may be physically different from the global states. Wilsky et.al. [7] have recently shown that a necessary and sufficient condition for our being able to recover globally optimal state estimates from locally optimal ones is that

$$C_{\mathbf{k}}^{\mathbf{i}} = H_{\mathbf{k}}^{\mathbf{i}} M_{\mathbf{k}}^{\mathbf{i}} \tag{12}$$

and for the tracking application, we expect that

$$x_{k}^{i} = M_{k}^{i} x_{k}$$
 (13)

where $M_{K}^{\hat{i}}$ is a matrix which results in the correct partitioning of global states to the subsystem filters. Alternatively, equation (12) allows us to define the local state vector in terms of the local coordinate system. In this case, $M_{K}^{\hat{i}}$ is the orthogonal matrix which defines the transformation between local and global coordinate systems.

^{*}Lower case variables are vector quantities while upper case generally corresponds to matrices of appropriate dimension. Also \mathbf{w}_k , \mathbf{x}_0 , and \mathbf{v}_k^1 through \mathbf{v}_k^M are uncorrelated with each other for all k and $N(\alpha,\Gamma)$ signifies an α mean white Gaussian process with covariance matrix Γ .

1.1 The Decentralized Square Root Information Filter as a Solution

Decentralized processing is achieved by distributing the minimization of the conventional least squares performance criterion amongst the local filters and global merging equations that follow (see [1] or our Phase I proposal to DARPA for details). Let

$$M_{k}^{i} = I \tag{14}$$

The best distribution for multisensor tracking is probably to minimize $| \mid \mid H_{K}^{i} \times k - z_{K}^{i} \mid \mid^{2}$ in each of the M local filters and minimize the remaining two terms in the performance criterion within the central (merge) processor. However, this point will be explored in detail in Phase II research. Thus, each of the tracking sensors may be processed using the standard SRIF mechanization. For the ith sensor, we have

Measurement Update

$${}^{1}T_{K}^{i} \begin{bmatrix} R_{K}^{i}(-) & z_{K}^{i}(-) \\ (R_{K}^{i})^{-\frac{1}{2}} H_{K}^{i} & (R_{K}^{i})^{-\frac{1}{2}} y_{K}^{i} \end{bmatrix} = \begin{bmatrix} R_{K}^{i}(+) & z_{K}^{i}(+) \\ 0 & e_{K}^{i} \end{bmatrix}$$
(15)

Time Update

$$2_{\mathsf{T}_{\mathsf{K}}^{\dot{\mathsf{I}}}} \begin{bmatrix} \mathsf{R}_{\mathsf{W}}(\mathsf{k}) & 0 & \mathsf{z}_{\mathsf{W}}(\mathsf{k}) \\ -\mathsf{R}_{\mathsf{K}}^{\dot{\mathsf{I}}}(+) & \mathbf{1}_{\mathsf{K}}^{-1} & \mathsf{R}_{\mathsf{K}}^{\dot{\mathsf{I}}}(+) & \mathbf{1}_{\mathsf{K}}^{-1} & \mathsf{z}_{\mathsf{K}}^{\dot{\mathsf{I}}}(+) \end{bmatrix}$$

$$= \begin{bmatrix} (i) & (i) & (i) \\ R_{w}^{*}(k) & R_{wx}^{*}(k) & z_{w}^{*}(k) \\ & & (i) & (i) \\ 0 & R_{k+1}(-) & z_{k+1}(-) \end{bmatrix}$$
(16)

where

$$[R_0^i(-) z_0^i(-)] = [0 \ 0],$$
 (17)

$$[R_{w}(k) \quad z_{w}(k)] = [0 \quad 0],$$
 (18)

the a priori estimate $\{R_0 (-) z_0(-)\}$ has covariance

$$-1$$
 $-T$ $P_0(-) = R_0(-) R_0(-)$, (19)

$$-1 -T$$
 $Q_{k} = R_{w} (k) R_{w} (k) , (20)$

 $z_{\rm W}(k)=R_{\rm W}(k)$ times the a priori expectation of $w_{\rm K}$, and the pair { $z_{\rm K}^i$, $H_{\rm K}^i$ } correspond to normalized measurement equations i.e., $R_{\rm K}^i=I$. Also, ${}^iT_{\rm K}^i$ and ${}^2T_{\rm K}^i$ are orthogonal transformations which put the matrices on the left hand sides of (15) and (16) respectively into upper triangular form. They may be implicitly computed using Householder transformations.

The local smoothing coefficients are combined with process noise and prior on \mathbf{x}_0 by solving the following recursive equation:

$$= \begin{bmatrix} R_{W}^{*}(k) & R_{WX}^{*}(k) & z_{W}^{*}(k) \\ 0 & H^{*}(k) & z_{W}^{*}(k) \\ 0 & 0 & \# \end{bmatrix}$$
 (21)

$$[H^{*}(-1) z^{*}(-1)] = [R_{0}(-) z_{0}(-)]$$
 (22)

and $H^*(k)$ is upper triangular. To obtain the globally optimal information vector $z_k(+)$ and square root information matrix $R_k(+)$, we solve the following equations using $H^*(k)$ and $z^*(k)$ from (21):

$${}^{4}T_{k} \begin{bmatrix} (1) & (1) \\ R_{k}(+) & z_{k}(+) \\ \vdots & \vdots \\ (M) & (M) \\ R_{k}(+) & z_{k}(+) \\ H^{*}(k-1) & z^{*}(k-1) \end{bmatrix} = \begin{bmatrix} R_{k}(+) & z_{k}(+) \\ 0 & \# \end{bmatrix}$$
(23)

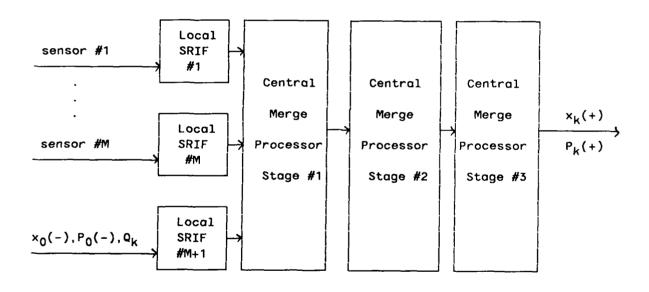
where $^4T_{\rm K}$ is an orthogonal transformation which puts the left hand side of (23) in upper triangular form. Globally optimal filter estimates and covariances are then given by

$$x_{k}(+) = R_{k}(+) z_{k}(+)$$
 (24)

$$-1$$
 -tr
 $P_{k}(+) = R_{k}(+) R_{k}(+)$ (25)

When the a priori information about the initial condition and process noise models are adjusted, it is only necessary to rerun (21) and (23) without reprocessing any measurements.

To summarize, the measurements from each tracking sensor may be processed by a local SRIF $\,$



(15)-(20) which generates a set of smoothing coefficients R_W^* , (i) (i) (i) $R_{WX}^*(+)$, $z_W^*(+)$ as well as a square root information matrix R(+) and

information vector z(+) . The central or merge processor consists of three separate processors which operate in parallel. The first mechanizes (21),(22) which combines the local smoothing coefficients with the effects of process noise and prior information about the initial state. The second mechanizes (23) which merges the local square root information matrices and vectors with output from the first, but only upon demand by the third. The third produces

estimates and covariances whenever desired by back-solving (24) and (25) respectively, noting that (24),(25) require output from the second processor. An important observation is that feedback of information from the merge processor to the local filters is not necessary here. This helps to keep the bandwidth of communication between the local processors and merge processor from exceeding hardware limitations.

2.0 Results of the Phase I Work

In order to validate our decentralized approach to solving the linear least squares estimation problem, two tasks were performed. First, in the next section 2.1 we construct and execute a simulation of the DSRIF within an underwater test range environment. The nominal trajectory is assumed to be known a priori so that actually, perturbations to the nominal are estimated. Then, in section 2.2 we complete our processing of real Multiple Rocket Launch System Data provided by the White Sands Missile Range. In this case, an adaptive form of the extended DSRIF, wherein the process noise levels were chosen to be a function of the globally optimal estimate error covariance, was used.

2.1 Validation of the DSRIF for a Simulated Underwater Zig-Zag Maneuver

A low level simulation of an underwater multisensor network, tracking a maneuvering underwater vehicle was encoded in Fortran '77 and executed on an IBM (clone) Model AT desktop computer (640K ram, 40 Mbyte hard disk, Intel 80287 math coprocessor). Initial conditions for the nominal trajectory were calculated using a "flat earth" or constant gravitational acceleration model which neglects the earth's rotation, neglects hydrodynamic drag, and assumes that the target vehicle is a point mass.

The initial position is located within the polygon formed by connecting adjacent sensors, and the desired terminal position is 1.86 miles downrange. However, 2 instantaneous reversals of the cross range velocity at 42 and 124 seconds were made, and this resulted in a total path length of 2.62 miles when projected onto the "Range Area General Map" (RAGM)" (which we use to define the GCS). That is, x^1, x^2, x^3 is a right handed coordinate system where the vector cross product of x^1 with x^2 is equal to x^3 , and the vectors $[x^1 \ 0 \ 0]$ and $[0 \ x^2 \ 0]$ point east along latitude 32.380 degrees and north along longitude 106.481 degrees respectively. Then $[0 \ 0 \ x^3]$ is collinear with the radial vector which points outward from the earth's center and passes through the origin of the GCS. Choosing the launch elevation angle, w.r.t. the x^1, x^2 tangent plane, to be -3.96 degrees results in a 200 ft depth variation over the path length. The corresponding initial position and velocity is

$$\begin{bmatrix} x_0^1 \\ x_0^2 \\ x_0^3 \\ x_0^6 \\ x_$$

and the total time to traverse the path was 167 seconds.

Five sensor locations which surround the projected flight path were selected. Each sensor records measurements with respect to its own LCS so that coordinate transformations to the GCS were derived and included in the observational equations. The transformation is

$$\begin{bmatrix} x^1 \\ x^2 \\ x^3 \end{bmatrix} = (T^i) \begin{bmatrix} x^{i,1} \\ x^{i,2} \\ x^{i,3} \end{bmatrix} + d^i$$
 (27)

where

$$T^{i} = (8T^{i})(7T^{i})(5T^{i})$$
 (28)

and

$$6_{T}i = \begin{bmatrix} \cos \beta^{i} & 0 & -\sin \beta^{i} \\ 0 & 1 & 0 \\ \sin \beta^{i} & 0 & \cos \beta^{i} \end{bmatrix}$$
 (30)

$$7_{T^{i}} =
\begin{bmatrix}
1 & 0 & 0 \\
0 & \cos L & -\sin L \\
0 & \sin L & \cos L
\end{bmatrix}$$
(31)

noting that $(T^i)^{tr} = (T^i)^{-1}$ since T^i is an orthogonal transformation. A computational savings results when $(^5T^i)(^6T^i)(^7T^i)$ are combined using trigonometric identities for the cosine and sine of the sum of two angles along with the small angle approximation since α^i is bounded by $\pm .025$ degrees over the length of the underwater Test Range.

 d^i Is the vector from the GCS to the ith LCS and α^i,β^i,τ^i are the three Euler angles describing the orientation of the ith LCS w.r.t. the GCS. Specifically, if we first rotate about x^1 counterclockwise by an $(L+\alpha^i)$ degree change in latitude (aligning y^2 with the polar axis), and then rotate counterclockwise about y^2 by a β^i degree change in longitude, (β^i is bounded by $\pm.025$ degrees over the width of the Test Range) then rotate clockwise about x^3 by an L degree change in latitude, and finally rotate about z^4 by τ^i degrees to account for "tangent plane misalignment", the LCS will coincide with the GCS when $d^i\approx 0$. α^i Is equal to the LCS latitude – GCS latitude. β^i Is equal to the LCS longitude – GCS longitude.

To determine $\alpha^{\dot{1}},~\beta^{\dot{1}}$ given L, § and $d^{\dot{1}}$ (in global coordinates), solve the following two equations:

$$\begin{bmatrix} ox^{i,1} \\ ox^{i,2} \\ ox^{i,3} \end{bmatrix} = \begin{bmatrix} ox^{1} \\ ox^{2} \\ ox^{3} \end{bmatrix} + T^{g} \begin{bmatrix} d^{i,1} \\ d^{i,2} \\ d^{i,3} \end{bmatrix}$$
GCS
$$(33)$$

where

$$\begin{bmatrix} x^{1} \\ ox^{2} \\ ox^{3} \\ ox^{3} \end{bmatrix} = \begin{bmatrix} r_{e} \cos L \cos \delta \\ -r_{e} \cos L \sin \delta \\ r_{e} \sin L \end{bmatrix}$$
 (34)

and

$$T^{g} = \begin{bmatrix} \sin \delta & -\cos \delta & 0 \\ \cos \delta & \sin \delta & 0 \\ 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} 1 & 0 & 0 \\ 0 & \sin L & -\cos L \\ 0 & \cos L & \sin L \end{bmatrix}$$
(35)

$$\begin{bmatrix} o_{\alpha_{i,2}}^{i,1} \\ o_{\alpha_{i,3}}^{i,2} \\ o_{\alpha_{i,3}}^{i,3} \end{bmatrix}_{\text{ECEF}} = \begin{bmatrix} o_{\alpha_{i}}^{i} \cos(L+\alpha_{i}^{i}) \cos(\delta+\beta_{i}^{i}) \\ o_{\alpha_{i}}^{i} \cos(L+\alpha_{i}^{i}) \sin(\delta+\beta_{i}^{i}) \\ o_{\alpha_{i}}^{i} \sin(L+\alpha_{i}^{i}) \end{bmatrix}$$
(36)

The solution is

$$L+\alpha^{i} = \tan^{-1} \frac{(o^{x^{i},3})^{2}}{(o^{x^{i},1})^{2} + (o^{x^{i},2})^{2}}$$
 (L+\alpha^{i}) \approx 32 degrees (37)

$$\delta + \beta^{i} = \cos^{-1} \frac{o^{x^{i}, 1}}{o^{r^{i}} \cos (L + \alpha^{i})}$$
 (\$\delta + \beta^{i}\$) \approx 106 degrees (38)

where

$$o^{r^{i}} = \frac{o^{x^{i},3}}{\sin(L+\alpha^{i})}$$
 (39)

The results are given in Table 1 below.

i	Sensor Type	d ^{i,1}	d ^{i,2}	d ^{i,3}	L+a ⁱ	δ+β ⁱ	γi
1	ps	0.50	0.50	0.	32.3872	106.4724	0.
2	ps	0.25	3.00	0.	32.4233	106.4767	0.
3	as	2.00	4.00	0.	32.4378	106.4467	0.
4	ps	2.50	2.00	Ο.	32.4089	106.4382	0.
5	as	3.00	1.00	0.	32.3944	106.4296	0.

Table 1: Passive Sonar (ps) and Active Sonar (as) Locations and Orientations in GCS for the Underwater Target

Detailed equations that describe the translational motion of the target were developed. The equations include a radial gravitational force as well as centrifugal and Coriolis forces which come about by rotation of the GCS about the polar axis. The equations are

$$\dot{x}^{4} = \frac{-\mu x^{1}}{((x^{1})^{2} + (x^{2})^{2} + (x^{3} + r_{e})^{2})^{3/2}} + 2\Omega(x^{5} \sin L - x^{6} \cos L) + \Omega^{2} x^{1}$$
 (40)

$$\dot{x}^{5} = \frac{-\mu x^{2}}{((x^{1})^{2} + (x^{2})^{2} + (x^{3} + r_{e})^{2})^{3/2}} - 2\Omega x^{4} \sin L + \Omega^{2}(x^{2} \sin^{2} L - (x^{3} + r_{e}) \cos L \sin L)$$
(41)

$$\dot{x}^{6} = \frac{-\mu(x^{3}+r_{e})}{((x^{1})^{2} + (x^{2})^{2} + (x^{3}+r_{e})^{2})^{3/2}} + 2\Omega x^{4}\cos L + \Omega^{2}((x^{3}+r_{e})\cos^{2}L - x^{2}\cos L \sin L) (42)$$

where

$$\mu = G m_{\Theta}$$
 (43)

and all other variables are defined in the List of Symbols. A spherical harmonic expansion of the earth's gravitational field is not necessary since the variation in target depth was small relative to the earth's radius. However, a major effect is hydrodynamic drag and buoyancy which we accounted for by reducing G to 1/1000 of its correct value. Thus, the trajectories tended to be unstable (due to centrifugal forces dominating) but reasonable ones could be obtained by careful tuning of the initial condition. More precise modeling will be undertaken in Phase II.

Any one of three sensor types may reside at each sensor location. The three types are

- active sonar wherein range (r^i), azimuth (Γ^i) and elevation (θ^i) data are available.
- · passive sonar wherein azimuth and elevation data are available, and
- doppler sonar wherein range, range rate $(\hat{r^i})$, azimuth and elevation data are available.

The observational equations in terms of the global state are

$$r^{i} = ((x^{1}-d^{i}, 1)^{2} + (x^{2}-d^{i}, 2)^{2} + (x^{3}-d^{i}, 3)^{2})^{1/2}$$
(44)

$$\theta^{i} = \sin^{-1} \frac{\begin{bmatrix} x^{1}-d^{i}, 1 \\ x^{2}-d^{i}, 2 \\ x^{3}-d^{i}, 3 \end{bmatrix}}{((x^{1}-d^{i}, 1)^{2} + (x^{2}-d^{i}, 2)^{2} + (x^{3}-d^{i}, 3)^{2})^{1/2}}$$
(45)

$$\Gamma^{i} = \tan^{-1} \frac{\begin{bmatrix} x^{1}-d^{i}, 1 \\ x^{2}-d^{i}, 2 \\ x^{3}-d^{i}, 3 \end{bmatrix}}{\begin{bmatrix} x^{1}-d^{i}, 1 \\ x^{2}-d^{i}, 2 \end{bmatrix}}$$

$$\begin{bmatrix} 1 & 0 & 0 \end{bmatrix} (T^{i})^{tr} \begin{bmatrix} x^{1}-d^{i}, 1 \\ x^{2}-d^{i}, 2 \\ x^{3}-d^{i}, 3 \end{bmatrix}$$
(46)

$$\begin{bmatrix}
x^{4} & x^{5} & x^{6}
\end{bmatrix} \begin{bmatrix}
x^{1}-d^{1}, 1 \\
x^{2}-d^{1}, 2 \\
x^{3}-d^{1}, 3
\end{bmatrix}$$

$$\dot{f}^{i} = \frac{((x^{1}-d^{1}, 1)^{2} + (x^{2}-d^{1}, 2)^{2} + (x^{3}-d^{1}, 3)^{2})^{1/2}}{((x^{1}-d^{1}, 1)^{2} + (x^{2}-d^{1}, 2)^{2} + (x^{3}-d^{1}, 3)^{2})^{1/2}}$$
(47)

The various measurement variables are defined by Figure 1 that follows.

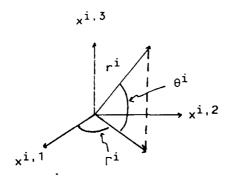


Figure 1: The Local Coordinate System and Measurement Variables

The nonlinear equations of motion (40)-(42) were integrated forwards in time starting from (26) and using the Fourth Order Runge Kutta method with a 1 second interval. Figures 2 and 3 show the resultant trajectory of the target vehicle. As a further check on the Fortran code for this part of the simulation, the total translational energy (kinetic plus potential) of the target was computed for each point along the computed trajectory. The total energy remained constant to within 1% as it should since there are no external forces (u(t) = 0) and the system is conservative. Figures 4 through 8 show some of the corresponding measurements for the various datatypes with $\mathbf{v}^{\dot{\mathbf{I}}} = \mathbf{0}$ for $\{1 \le \mathbf{i} \le \mathbf{M}\}$.

Each row of A(t) is computed by partial differentiating the same row of f(x(t)) w.r.t. the nominal state. Thus, A(t) has the following structure

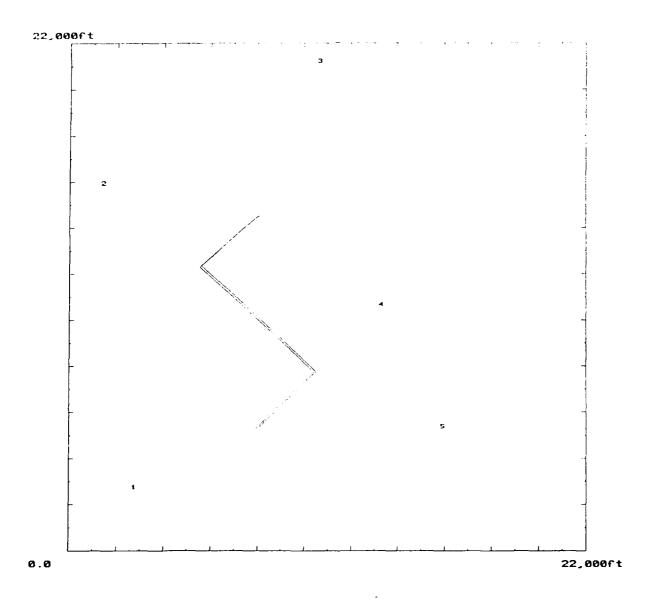


Figure 2

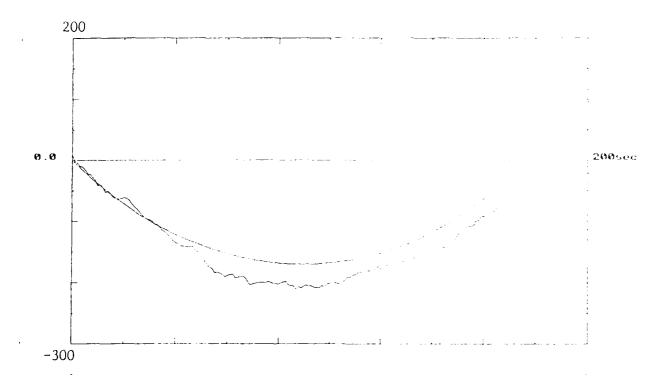


Figure 3

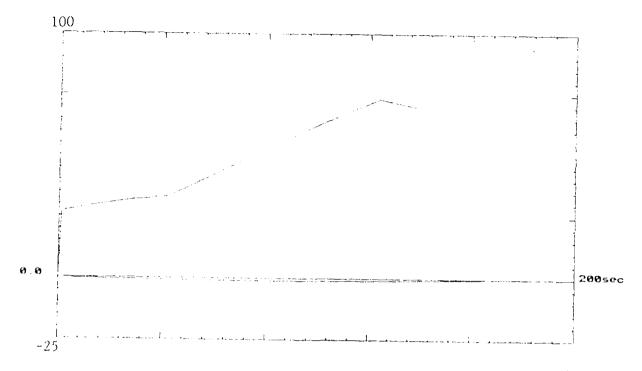


Figure 4

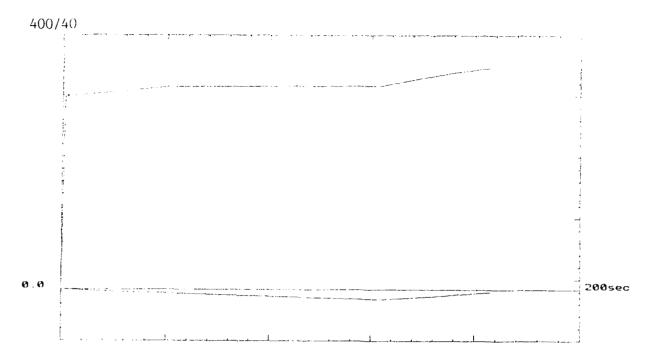


Figure 5

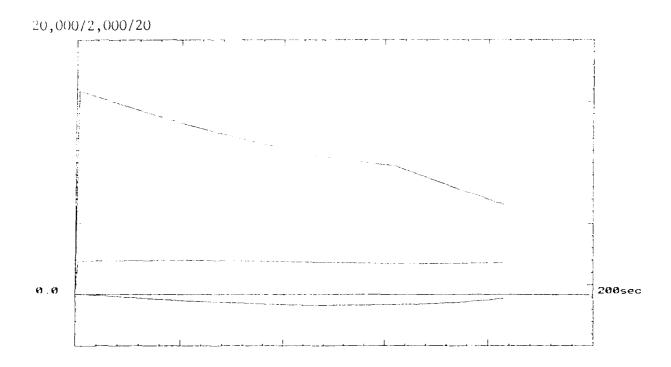


Figure 6

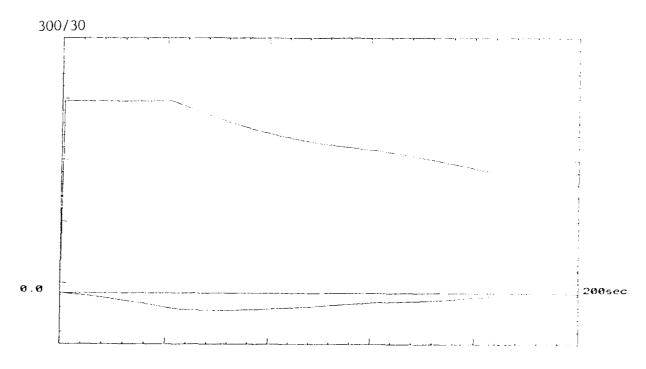


Figure 7

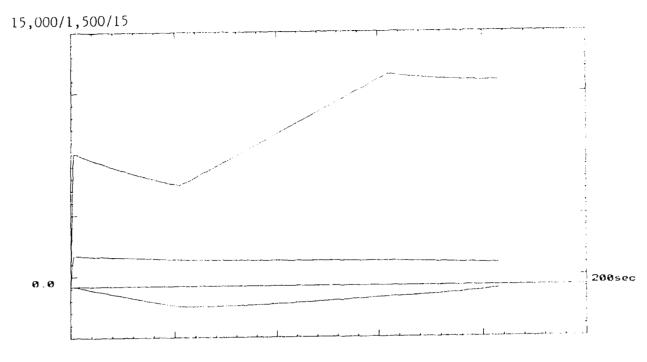


Figure 8

where

$$\frac{3x^{4}}{3x^{1}} = \frac{-\mu}{((x^{1})^{2} + (x^{2})^{2} + (x^{3}+r_{e})^{2})^{3/2}} + \frac{3\mu(x^{1})^{2}}{((x^{1})^{2} + (x^{2})^{2} + (x^{3}+r_{e})^{2})^{5/2}} + \frac{\Omega^{2}}{(49)}$$

$$\frac{3 \dot{x}^4}{3 x^2} = \frac{3 \mu x^1 x^2}{((x^1)^2 + (x^2)^2 + (x^3 + r_e)^2)^{5/2}}$$
 (50)

$$\frac{3x^{4}}{x^{3}} = \frac{3\mu x^{1}(x^{3}+r_{e})}{((x^{1})^{2} + (x^{2})^{2} + (x^{3}+r_{e})^{2})^{5/2}}$$
(51)

$$\frac{3x^4}{3x^4} = 0 \tag{52}$$

$$\frac{3 x^4}{-x^5} = 2\Omega \sin L \tag{53}$$

$$\frac{3 \cdot x^4}{-x^6} = -2\Omega \cos L \tag{54}$$

$$\frac{3 \cdot x^{5}}{3 \cdot x^{1}} = \frac{3 \mu x^{1} x^{2}}{((x^{1})^{2} + (x^{2})^{2} + (x^{3} + r_{e})^{2})^{5/2}}$$
(55)

$$\frac{3 \dot{x}^{5}}{3 x^{2}} = \frac{-\mu}{((x^{1})^{2} + (x^{2})^{2} + (x^{3} + r_{e})^{2})^{3/2}} + \frac{3\mu(x^{2})^{2}}{((x^{1})^{2} + (x^{2})^{2} + (x^{3} + r_{e})^{2})^{5/2}} + \Omega^{2} \sin^{2} L$$
(56)

$$\frac{3x^{5}}{3x^{3}} = \frac{3\mu x^{2}(x^{3}+r_{e})}{((x^{1})^{2} + (x^{2})^{2} + (x^{3}+r_{e})^{2})^{5/2}} - \Omega^{2}\cos L \sin L$$
 (57)

$$\frac{3 \times^5}{\lambda^4} = -2\Omega \sin L \tag{58}$$

$$\frac{\int \dot{x}^5}{\int x^5} = 0 \tag{59}$$

$$\frac{\int \dot{x}^5}{\lambda^6} = 0 \tag{60}$$

$$\frac{\int \dot{x}^{6}}{\int x^{1}} = \frac{3\mu x^{1}(x^{3}+r_{e})}{((x^{1})^{2} + (x^{2})^{2} + (x^{3}+r_{e})^{2})^{5/2}}$$
(61)

$$\frac{\int \dot{x}^6}{\int x^2} = \frac{3\mu x^2 (x^3 + r_e)}{((x^1)^2 + (x^2)^2 + (x^3 + r_e)^2)^{5/2}} - \Omega^2 \cos L \sin L$$
 (62)

$$\frac{\int \dot{x}^{6}}{\int x^{3}} = \frac{-\mu}{((x^{1})^{2} + (x^{2})^{2} + (x^{3}+r_{e})^{2})^{3/2}} + \frac{3\mu(x^{3}+r_{e})^{2}}{((x^{1})^{2} + (x^{2})^{2} + (x^{3}+r_{e})^{2})^{5/2}} + \Omega^{2}\cos^{2}L$$
(63)

$$\frac{\int \dot{x}^6}{\sqrt{x^4}} = 2\Omega \cos L \tag{64}$$

$$\frac{\int \dot{x}^6}{\sqrt{x^5}} = 0 \tag{65}$$

$$\frac{7 \times_{e}}{7 \times_{e}} = 0 \tag{66}$$

Each row of $C^i(t)$ is computed by partial differentiating the same row of $hi(x(t) \ w.r.t.$ the nominal state. Thus, when the ith sensor is a doppler sonar, $C^i(t)$ has the following structure

$$C^{i}(t) = \begin{bmatrix} \frac{1}{3}r^{i} & & \frac{1}{3}r$$

where

$$\frac{\int r^{i}}{\int_{x^{1}}} = \frac{x^{1}-d^{i,1}}{((x^{1}-d^{i,1})^{2} + (x^{2}-d^{i,2})^{2} + (x^{3}-d^{i,3})^{2})^{1/2}}$$
(68)

$$\frac{\int r^{i}}{\int x^{2}} = \frac{x^{2} - d^{i,2}}{((x^{1} - d^{i,1})^{2} + (x^{2} - d^{i,2})^{2} + (x^{3} - d^{i,3})^{2})^{1/2}}$$
(69)

$$\frac{\int r^{i}}{\int x^{3}} = \frac{x^{3}-d^{i}, 3}{((x^{1}-d^{i}, 1)^{2} + (x^{2}-d^{i}, 2)^{2} + (x^{3}-d^{i}, 3)^{2})^{1/2}}$$
(70)

$$\frac{3ri}{3x^4} = \frac{3ri}{3x^5} = \frac{3ri}{3x^6} = 0 \tag{71}$$

$$\frac{\sqrt[3]{r_i}}{\sqrt[3]{x^1}} = -\frac{(x^4(x^1-d^{i,1}) + x^5(x^2-d^{i,2}) + x^6(x^3-d^{i,3}))(x^1-d^{i,1})}{((x^1-d^{i,1})^2 + (x^2-d^{i,2})^2 + (x^3-d^{i,3})^2)^{3/2}}$$
(72)

$$+\frac{x^{4}}{((x^{1}-d^{1},1)^{2}+(x^{2}-d^{1},2)^{2}+(x^{3}-d^{1},3)^{2})^{1/2}}$$
(73)

$$\frac{\sum_{x^2} \dot{r}i}{\sum_{x^2}} = -\frac{(x^4(x^1 - d^{\dot{1}, 1}) + x^5(x^2 - d^{\dot{1}, 2}) + x^6(x^3 - d^{\dot{1}, 3}))(x^2 - d^{\dot{1}, 2})}{((x^1 - d^{\dot{1}, 1})^2 + (x^2 - d^{\dot{1}, 2})^2 + (x^3 - d^{\dot{1}, 3})^2)^{3/2}} + \frac{x^5}{((x^1 - d^{\dot{1}, 1})^2 + (x^2 - d^{\dot{1}, 2})^2 + (x^3 - d^{\dot{1}, 3})^2)^{1/2}}$$
(74)

$$\frac{\sqrt[3]{x^3}}{\sqrt[3]{x^3}} = -\frac{(x^4(x^1-d^{1,1}) + x^5(x^2-d^{1,2}) + x^6(x^3-d^{1,3}))(x^3-d^{1,3})}{((x^1-d^{1,1})^2 + (x^2-d^{1,2})^2 + (x^3-d^{1,3})^2)^{3/2}}$$

$$+\frac{x^{6}}{((x^{1}-d^{1},1)^{2}+(x^{2}-d^{1},2)^{2}+(x^{3}-d^{1},3)^{2})^{1/2}}$$
(75)

$$\frac{\sqrt[3]{r^i}}{\sqrt[3]{x^4}} = \frac{x^1 - d^{i,1}}{((x^1 - d^{i,1})^2 + (x^2 - d^{i,2})^2 + (x^3 - d^{i,3})^2)^{1/2}}$$
(76)

$$\frac{\sqrt[3]{r^{i}}}{\sqrt[3]{x^{5}}} = \frac{x^{2}-d^{i}, 2}{((x^{1}-d^{i}, 1)^{2} + (x^{2}-d^{i}, 2)^{2} + (x^{5}-d^{i}, 3)^{2})^{1/2}}$$
(77)

$$\frac{\int_{x^{6}}^{\dot{r}^{i}}}{\int_{x^{6}}^{x^{6}}} = \frac{x^{3-d^{i},3}}{((x^{1-d^{i},1})^{2} + (x^{2-d^{i},2})^{2} + (x^{3-d^{i},3})^{2})^{1/2}}$$
(78)

$$\frac{\int \theta^{i}}{\int x^{1}} = \frac{1}{(1-u^{2})^{1/2}} - \frac{\left[0\ 0\ 1\right] \left(T^{i}\right)^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3 \end{bmatrix}}{((x^{1}-d^{i},1)^{2} + (x^{2}-d^{i},2)^{2} + (x^{3}-d^{i},3)^{2})^{3/2}}$$

$$+ \frac{\left[0 \ 0 \ 1\right] (T^{i})^{tr} \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}}{((x^{1}-d^{i}, 1)^{2} + (x^{2}-d^{i}, 2)^{2} + (x^{3}-d^{i}, 3)^{2})^{1/2}}$$
(79)

where u is the argument of \sin^{-1} in equation (45).

$$\frac{\int e^{i}}{\int x^{2}} = \frac{1}{(1-u^{2})^{1/2}} - \frac{[0 \ 0 \ 1] \ (7^{i})^{tr} \begin{bmatrix} x^{1}_{2}-d^{i}, 1 \\ x^{2}_{2}-d^{i}, 2 \\ x^{3}_{3}-d^{i}, 3 \end{bmatrix}}{(x^{2}_{3}-d^{i}, 3)^{2}} (x^{2}_{3}-d^{i}, 3)^{2} (x^{2}_{3}-d^{i}, 3)^{2}} \\
= \begin{bmatrix} [0 \ 0 \ 1] \ (7^{i})^{tr} \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix} \\
+ \frac{((x^{1}_{3}-d^{i}, 1)^{2} + (x^{2}_{3}-d^{i}, 2)^{2} + (x^{3}_{3}-d^{i}, 3)^{2})^{1/2}}{((x^{1}_{3}-d^{i}, 1)^{2} + (x^{2}_{3}-d^{i}, 2)^{2} + (x^{3}_{3}-d^{i}, 3)^{2})^{1/2}} \\
+ \frac{\int e^{i}}{((x^{1}_{3}-d^{i}, 1)^{2} + (x^{2}_{3}-d^{i}, 2)^{2} + (x^{3}_{3}-d^{i}, 3)^{2})^{1/2}} \\
+ \frac{\int e^{i}}{((x^{1}_{3}-d^{i}, 1)^{2} + (x^{2}_{3}-d^{i}, 2)^{2} + (x^{3}_{3}-d^{i}, 3)^{2})^{1/2}} \\
+ \frac{\int e^{i}}{(1+u^{2})} - \frac{\int e^{i}}{(1+u^{2})} - \frac{\int e^{i}}{(1+u^{2})^{2}} \\
- \frac{[0 \ 1 \ 0 \ 1] \ (7^{i})^{tr} \begin{bmatrix} x^{1}_{3}-d^{i}, 1 \\ x^{2}_{3}-d^{i}, 3} \\ x^{3}_{3}-d^{i}, 3} \end{bmatrix} [1 \ 0 \ 0] \ (7^{i})^{tr} \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}} \\
- \frac{\int e^{i}}{(1+u^{2})^{2}} - \frac{1}{(1+u^{2})^{2}} \\
- \frac{[0 \ 1 \ 0] \ (7^{i})^{tr} \begin{bmatrix} x^{1}_{3}-d^{i}, 1 \\ x^{2}_{3}-d^{i}, 3} \\ x^{3}_{3}-d^{i}, 3} \end{bmatrix} [1 \ 0 \ 0] \ (7^{i})^{tr} \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}} \\
- \frac{\int e^{i}}{(1+u^{2})^{2}} - \frac{1}{(1+u^{2})^{2}} \\
- \frac{(1 \ 0 \ 0] \ (7^{i})^{tr} \begin{bmatrix} x^{1}_{3}-d^{i}, 1 \\ x^{2}_{3}-d^{i}, 3} \\ x^{3}_{3}-d^{i}, 3} \end{bmatrix} [1 \ 0 \ 0] \ (7^{i})^{tr} \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}} \\
- \frac{(1 \ 0 \ 0] \ (7^{i})^{tr} \begin{bmatrix} x^{1}_{3}-d^{i}, 1 \\ x^{2}_{3}-d^{i}, 3} \\ x^{3}_{3}-d^{i}, 3} \end{bmatrix} [1 \ 0 \ 0] \ (7^{i})^{tr} \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}} \\
- \frac{(1 \ 0) \ (7^{i})^{tr} \begin{bmatrix} x^{1}_{3}-d^{i}, 1 \\ x^{2}_{3}-d^{i}, 3} \\ x^{3}_{3}-d^{i}, 3} \end{bmatrix} [1 \ 0 \ 0] \ (7^{i})^{tr} \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}} \\
- \frac{(1 \ 0) \ (7^{i})^{tr} \begin{bmatrix} x^{1}_{3}-d^{i}, 1 \\ x^{2}_{3}-d^{i}, 3} \end{bmatrix}}{(1 \ 0 \ 0] \ (7^{i})^{tr} \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}}$$

$$[0 \ 1 \ 0] \ (T^{i})^{tr} \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix} + \frac{x^{1}-d^{i}, 1}{x^{2}-d^{i}, 2}$$

$$[1 \ 0 \ 0] \ (T^{i})^{tr} \begin{bmatrix} x^{1}-d^{i}, 1 \\ x^{2}-d^{i}, 2 \\ x^{3}-d^{i}, 3 \end{bmatrix}$$
(83)

where u is now the argument of tan^{-1} in equation (46).

$$\frac{\int \Gamma^{i}}{\int x^{2}} = \frac{1}{(1+u^{2})} - \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3 \end{bmatrix}}{\left(\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3 \end{bmatrix}} \right)^{2}} \\
+ \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} 0\\ 1\\ 0 \end{bmatrix}}{\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3 \end{bmatrix}} \\
- \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3 \end{bmatrix}}{\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} 0\\ 0\\ 1\\ 1 \end{bmatrix}} \\
- \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3 \end{bmatrix}}{\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} 0\\ 0\\ 1\\ 1 \end{bmatrix}} \\
- \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3 \end{bmatrix}}{\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} 0\\ 0\\ 1\\ 1 \end{bmatrix}} \\
- \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3} \end{bmatrix}}{\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} 0\\ 0\\ 1\\ 1 \end{bmatrix}} \\
- \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3} \end{bmatrix}}{\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} 0\\ 0\\ 1\\ 1 \end{bmatrix}} \\
- \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3} \end{bmatrix}}{\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} 0\\ 0\\ 1\\ 1 \end{bmatrix}} \\
- \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3} \end{bmatrix}}{\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} 0\\ 0\\ 1\\ 1 \end{bmatrix}} \\
- \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3} \end{bmatrix}}{\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} 0\\ 0\\ 0\\ 1 \end{bmatrix}} \\
- \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3} \end{bmatrix}}{\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} 0\\ 0\\ 0\\ 1 \end{bmatrix}} \\
- \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3} \end{bmatrix}}{\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} 0\\ 0\\ 0\\ 1 \end{bmatrix}} \\
- \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{2}-d^{i},2\\ x^{3}-d^{i},3} \end{bmatrix}}{\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} 0\\ 0\\ 1\\ 1 \end{bmatrix}} \\
- \frac{\left[0\ 1\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{2}-d^{i},3} \end{bmatrix}}{\left[1\ 0\ 0\right] (\tau^{i})^{tr} \begin{bmatrix} x^{1}-d^{i},1\\ x^{2}-d^{i},2\\ x^{2}-d^{i},3} \end{bmatrix}}$$

$$[0 \ 1 \ 0] \ (T^{i})^{tr} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} + \frac{1}{x^{1}-d^{i}, 1} \begin{bmatrix} x^{1}-d^{i}, 1 \\ x^{2}-d^{i}, 2 \\ x^{3}-d^{i}, 3 \end{bmatrix}$$

$$(85)$$

$$\frac{3\Gamma^{i}}{3x^{4}} = \frac{3\Gamma^{i}}{3x^{5}} = \frac{3\Gamma^{i}}{3x^{6}} = 0 \tag{86}$$

The 6 biases $\alpha^i, \beta^i, \tau^i, d^{i,1}, d^{i,2}, d^{i,3}$ per local system could be included as states in the filter and estimated in order to correct for any preflight miscalibration. Then, $\alpha^i = 0$, $\beta^i = 0$... would be added to the equations of motion and additional partial derivative expressions would be needed. However, only the 6 positions and velocities of the underwater target were included.

In order to create simulated data and test the state estimation part of the code, a subroutine RANDOM for generating sequences of Gaussian random vectors with prescribed covariance was used. Two algorithms were considered in the derivation of the subroutine. The first proceeds by rotating coordinates to a system in which the covariance matrix is diagonal. In this system the multivariate normal density becomes equal to the product of its marginal densities, and each marginal density can be sampled independently of the other components. After obtaining a sample vector in this rotated system, the coordinates are rotated back to the original system.

The second algorithm proceeds by decomposing the multivariate normal density into the product of the marginal density of the first variate times the joint density of the remaining variates, conditional upon the value sampled for the first. This joint density is determined once the first variate has been sampled from its marginal density. The procedure is then applied to the second variate and iterated until values have been assigned to all components of the sample vector. This "Conditional Decomposition Algorithm" will execute more rapidly than the latter "Matrix Diagonalization Algorithm" especially for time varying covariance matrices. Thus it was chosen as the basis for subroutine RANDOM.

Assuming a constant covariance matrix, RANDOM has been tested by counting the number of random values within several bands for each component.

Comparison with theory has shown agreement to within a few percent.

Figures 9 and 10 show the evolution of the perturbed state as governed by equation (3). The initial condition is

 $\delta x_0 = [79.2ft \ 0.ft \ 0.ft/s \ 0.ft/s \ 0.ft/s] ^{tr}$

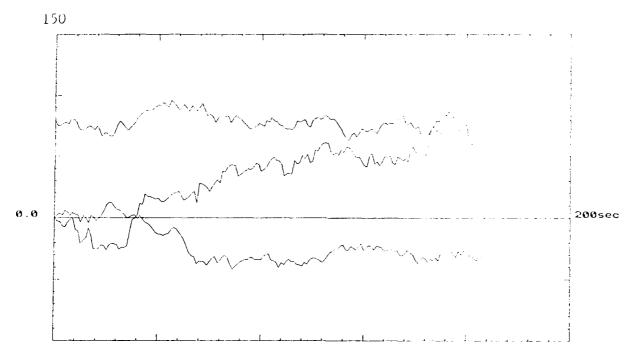
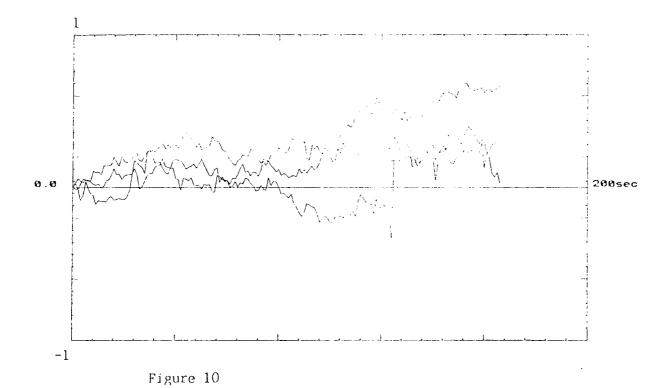


Figure 9



which represents a perturbation in the x location only. However, the resulting perturbed trajectories in the y and z directions are nonzero due to cross coupling terms within A(t). A diagonal Q_{k} with small variances was used to generate the process noise sequence. The position variances were 10 ft² and the velocity variances were 10^{-3} (ft/s)².

Figures 11, 12 and 13 are the corresponding DSRIF results with all of the prior and process noise information embedded in the merge processor. Each of the 12 local filters processed 1 measurement variable. A diagonal $\rm R_k$, with variances of $10^{-8}~\rm deg^2$ and 1 ft² for angular variables and range variables respectively, was used to generate the measurement noise sequence. The initial state estimate for all of the local filters was

$$\delta x_0^i = [179.2 \text{ft} \ 100. \text{ft} \ 10. \text{ft} \ 5.97 \text{ ft/s} \ 5.97 \text{ ft/s} \ -.41 \text{ ft/s}]^{tr}$$

and a diagonal $P_0(+)$, with variances of 1 ft² and 1 (ft/s)² for position variables and velocity variables respectively, was used to initialize the merge processor. Figures 11 and 12 show that the rms position estimate errors quickly decay to steady state mean values after only a few time samples however, the velocity estimate errors reach steady state after approximately 40 samples. The corresponding estimate error covariances in Figure 13 follow the same course as expected.

In Figure 14 the process noise levels were multiplied by 5 and 100 for position and velocity variances respectively. Comparison with Figure 13 shows that the corresponding estimate error covariances increase as well. This is as expected since $\mathbf{Q}_{\mathbf{k}}$ is linearly related to the time updated estimate error covariance i.e., the conventional covariance time update equation is given by

$$P_{k+1}(-) = \Phi_k P_k(+) \Phi_k^{tr} + Q_k$$
 (87)

Furthermore, the same phenomenon results when R_k is multiplied by a factor of 10⁸ for angular variables and 10 for range variables in Figure 15. The conventional covariance measurement update equation in Josephson Stabilized form

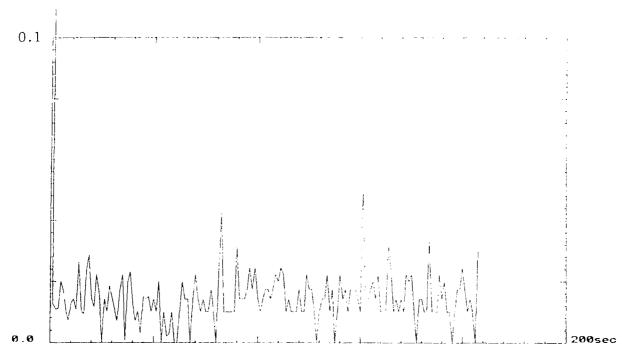
$$P_{k}(+) = [I - K_{k}H_{k}] P_{k}(-) [I - K_{k}H_{k}]^{tr} + K_{k}R_{k}K_{k}^{tr}$$
 (88)

may be combined with the Kalman gain equation

$$K_k = P_k(-)H_k^{tr} [H_k P_k(-)H_k^{tr} + R_k]^{-1}$$
 (89)

to show that the time updated estimate error covariance is linearly related to $R_{\bf k}$ as well. The result is that

$$P_{k}^{-1}(+) = P_{k}^{-1}(-) + H_{k}^{tr} R_{k}^{-1} H_{k}$$
 (90)



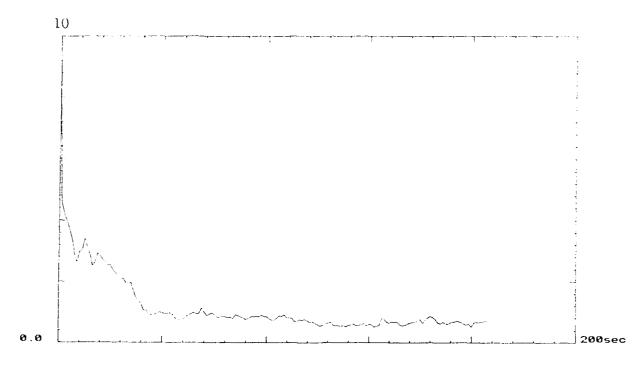


Figure 12

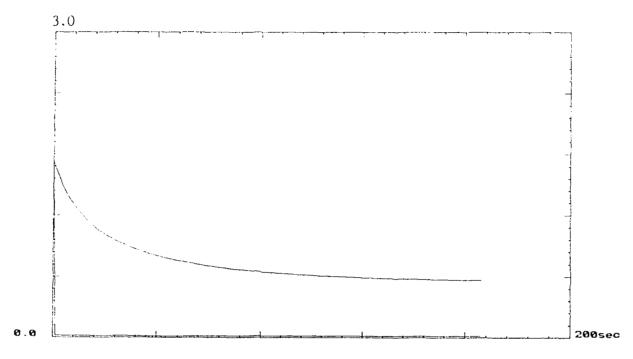


Figure 13

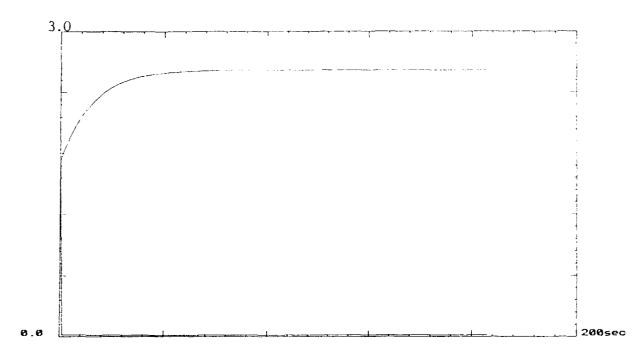


Figure 14

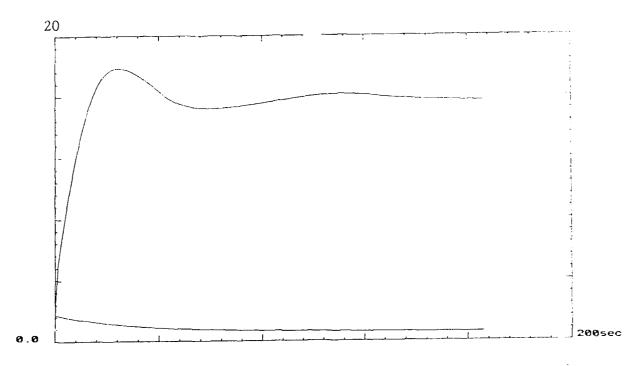


Figure 15

2.2 Extended-Decentralized Square Root Information Filtering of MLRS Data

Under SBIR Phase I contract DAAD07-87-C-0103 with the U.S. Army, White Sands Missile Range (WSMR), MTI previously derived an extended version of the algorithm and successfully used it to track real Multiple Launch Rocket System (MLRS) data provided by WSMR. In this section we describe some of the results obtained along with points of departure for future work that was needed. Then, in the next section 2.2.1 we finish processing the MLRS data.

On November 11, 1987 six rockets were launched sequentially in time over a period of 2 hours and 30 minutes at WSMR. Only 1 rocket was airborne at any one time and thus data association for multitarget tracking was not needed. The digitized measurements for all six shots were plotted in order to select the best set as characterized by the least amount of data drop outs and outlyers.

The MLRS data set contained azimuth and elevation angle measurements (with respect to each local sensor) from 11 optical trackers (analogous to passive hydrophones). The origin of the White Sands Coordinate System is at 32.38 degrees latitude and 106.481 degrees longitude. The data set also contained range, azimuth and elevation angle measurements from 3 radars (analogous to active sonar) but with respect to the local coordinate system originating at the launcher. The precise location and orientation of each sensor was known a priori and no attempt was made to estimate its uncertainty.

In this case, an extended version of the DSRIF is needed since the nominal rocket trajectories were unavailable. The same is true for underwater vehicle tracking although the velocities and accelerations are much smaller. The Extended DSRIF (E-DSRIF) may be derived by extending the observations equation, linearized about the current estimate, to

$$y_k^i = H_k^i x_k + v_k^i + z_k^i$$
 (91)

where

$$z_k^i = h^i(x) \Big|_{x=x_k(-)} - H_k^i x_k(-)$$
 (92)

and

$$H_{k}^{i} = \frac{\int h^{i}(x)}{\int x} \Big|_{x=x_{k}(-)}$$
 (93)

and extending the dynamics equation, linearized about the current estimate, to

$$x_{k+1} = F_k x_k + B_k w_k + g_k$$
 (94)

where

$$g_k = f(x) \Big|_{x=x_k(+)} - F_k x_k(+)$$
 (95)

and

$$F_{k} = \frac{\int x}{\int x}$$
 (96)

Substituting equations (91) and (92) into the least squares performance functional of equation (13) in our DARPA Phase I proposal gives the result (details were provided in [2]).

Processing on the global scale is the same as for the DSRIF i.e., the merge steps are exactly as defined in the Phase I proposal. Only processing on the local scale is different. A major difference between the E-DSRIF and DSRIF is that the local E-DSRIFilters require knowledge of the globally optimal estimate $\mathbf{x}_k(\underline{+})$ in order to compute their first order Taylor series expansion terms \mathbf{F}_k , \mathbf{g}_k , H_k^i and z_k^i whereas the DSRIF may compute $\mathbf{x}_k(\underline{+})$ at any rate less than the highest data rate. Future research should examine whether an E-DSRIF algorithm, in which the Taylor series expansions are about the locally optimal estimates, may be derived.

In order to derive a suitable dynamical model as well as initialize the filter, the position, velocity, acceleration and jerk of the rocket were precomputed using finite differencing with $\Delta t = .1$ seconds. Results were plotted in Figures 29 through 30 (of [2]) using all of the data provided for radar #350 except for the first 21 samples (we estimated that all radar trackers or rt's had locked onto the target by the 22nd sample). Figure 30 indicated that jerk could be suitably modeled as a white Gaussian noise process with constant mean. Thus, the E-DSRIF was encoded in Fortran '77 using a second order polynomial dynamical model.

Figures 31 through 35 showed the rt and optical tracker (ot) measurements as predicted by the E-DSRIF. Comparison with the actual measurements in Figures 16, 18, 19, 20 and 22 showed an exact match to within a plotting line width. A better means of comparison is thus provided below in Tables 2 and 3 where r_{k}^{i} is range in feet Γ_{k}^{i} is azimuth in degrees, and θ_{k}^{i} is elevation in degrees. Also, {m,e} corresponds to {actual measurements}, estimated measurements}.

The large values of Rk(j,j) for ot variables serves to weight the rt data much more heavily in computing estimates. Decreasing the ot measurement errors to more realistic values should give similar results since the predicted ot measurements matches their actual values very closely.

k	type	∍ r _k 12	ր <mark>12</mark>	θ <mark>12</mark>	r _k 14	Г <mark>1</mark> 4	θ <mark>14</mark>
23	m	4,355.61	359.18	18.07	4,305.91	357.65	18.04
	е	4,355.61	359.19	18.74	4,355.61	359.19	18.74
24	m	4.683.17	358.91	18.07	4,665.94	358.30	18.09
	e	4,658.07	358.16	18.07	4,658.07	358.16	18.07
229	m	52,171.33	358.96	11.88	52,163.32	358.95	11.89
	е	52,167.32	358.96	11.89	52,167.32	358.96	11.89
230	m	52,332.09	358.96	11.84	52,324.25	358.95	11.86
	е	52,328.15	358.96	11.85	52,328.15	358.96	11.85
:							
459	m	80,328.90	359.22	2.96	80,178.71	359.22	3.00
	е	80,254.10	359.22	2.98	80,254.10	359.22	2.98

Table 2: Radar #350 and #394 Measurements and Estimated Measurements for MLRS. $R_k(j,j)=10^{10}$ for ot variables and 10.,1.,1. for rt range, azimuth and elevation variables respectively. $Q_k(j,j)=$ diag [1.7 x 10¹⁷ 7.4 x 10¹⁷ 9.3 x 10¹⁷] .

k	type	Г <mark>3</mark>	θ <mark>3</mark>	Г <mark>4</mark>	θ <mark>4</mark>	Γ <mark>1</mark>	θ <mark>1</mark>
23	m	154.96	1.05	151.09	3.29	257.38	3.80
	е	154.97	. 18	151.87	2.01	255.49	2.88
24	m	154.39	1.19	150.73	3.58	258.16	4.07
	е	154.90	. 25	151.71	2.14	256.30	3.00
							!
229	m	68.57	28.54	22.78	21.75	***	***
	е	69.20	26.13	22.93	20.28	331.64	12.52
230	m	***	***	22.64	21.65	***	***
	e	68.74	26.07	22.80	20.18	331.73	12.48
459	m	26.07	5.58	***	***	***	***
	е	26.32	3.94	10.98	3.66	342.28	2.98

Table 3: G30, G80, G110 Measurements and Estimated Measurements for MLRS. $R_k(j,j)$ =10¹⁰ for ot variables and 10.,1.,1. for rt range, azimuth and elevation variables respectively. $Q_k(j,j)$ = diag [1.7 x 10¹⁷ 7.4 x 10¹⁷ 9.3 x 10¹⁷] .

Figures 36 and 37 showed the global position estimates and corresponding estimate error covariances respectively. Again, the rocket positions derived from radar #350 as compared with the E-DSRIF estimates based upon all of the 5 selected sensors, showed extremely close agreement. The slight difference in the estimate of height is due to using $T^{lauch} = I$ instead of its correct value as defined by equation (34) from [2]. In Table 4 below, the estimates are compared using the correct coordinate transformation.

k	tp	x _k ¹	ײk	x _k ³	× _k	x _k ⁵	×6 k	× ⁷ _k	x _k 8	×k ⁹
22	е -	14,104.	-243,740.	3,754.	-296.	3,110.	1,015.	430.	-23.	73.
	m same									
23	е -	14,189.	-243,437.	3,798.	-280.	3,104.	995.	429.	-23.	73.
	m -	14,076.	-243,413.	3,806.	-254.	3,120.	986.	436.	-23.	80.
24	e -	14,184.	-243,110.	3,903.	106.	3,271.	1,062.	497.	11.	85.
	m -	14,160.	-243,101.	3,905.	-211.	3,118.	994.	421.	-35.	49.
229	e -	15,018.	-196,406.	12,646.	-19.	1,639.	1.	4	-45.	-15.
	m -	15,012.	-196,397.	12,641.	-17.			1.6	-46.	-10.
230	e -	15.020.	-196,242.	12.646.	-20.	1,635.	-3.	-5.	-41.	-28.
			-196,233.		-18.		-4.	1.7	-49.	8.
•										
•										
459	A -	15. 197.	~167,388.	5,729.	22.	758.	-431.	21	-221.	184
700		•	~167,307.	•	8.	866.	-439.		-25.	

Table 4: Global Position Estimates and Derived Measurements for MLRS. $R_k(j,j)=10^{10}$ for ot variables and 10.,1.,1. for rt range, azimuth and elevation variables respectively. $Q_k(j,j) = \text{diag} \left[1.7 \times 10^{17} \right] 7.4 \times 10^{17} = 9.3 \times 10^{17}$.

Finally, the monotonically increasing estimate error covariance (actually, P_0 = diag [10^{12} ft² ... 10^{12} ft²/sec² ... 10^{13} ft²/sec⁴ ...] was used to initialize the covariance propagation so that the first step is a large, off scale decrease to approximately 100 ft²) is due to our using values of Q_k approximately 12 orders of magnitude higher than its precomputed sample value. A more realistic value should result in a P_k with quite the opposite behavior.

2.2.1 New Results

The E-DSRIF was successfully used to track an MLRS data set in previous work however, a monotonically increasing estimate error covariance was observed (see Figure 37 of [2]). This is due to our using values of \mathbf{Q}_k many orders of magnitude higher than its precomputed sample value in order to compensate for any errors in the model. For this subtask, \mathbf{Q}_k was adjusted interactively until an acceptable covariance function was obtained. As Figures A through C indicate, more realistic values of \mathbf{Q}_k results in a \mathbf{P}_k with quite the opposite (and more desirable) behavior previously illustrated.

 ${\bf Q_k}$ starts high at 10³ in Figures A through C and decreases monotonically to 10^{-18} . Both the position and velocity estimate error variances are still unstable when ${\bf Q_k}=10^3$ however both reach stability when ${\bf Q_k}=10^{-2}$. Not much improvement is gained by decreasing ${\bf Q_k}$ beyond 10^{-3} . One of the measurement variables from each of the 5 sensors selected were also plotted (not shown here) in Figures A through C. This was done in order to see whether the oscillations in estimate error variances, which begin to occur when ${\bf Q_k}=10^{-2}$, are correlated with the time intervals of data drop-out.

Next, the ot measurement errors were kept artificially high in Phase I. These large values of R_k for ot variables serves to weight the rt data much more heavily in computing estimates. In Figures D through F (where $Q_k\!=\!10^{-1}$) and G through I (where $Q_k\!=\!10^{-3}$), R_k was decreased for 3 of the 6 ot measurement variables. Decreasing some of the ot measurement errors from 10^{10} to more realistic values near 1 improved the results significantly by increasing the rate of convergence, especially near the initial time point.

Finally, an adaptive form of the E-DSRIF wherein

$$Q_{k} = factor * P_{k}(+)$$
 (97)

was encoded. This simple method for tuning the filter by using a feedback loop to compute Q_k , gave exceedingly good results as shown in Figures J through L (where all measurement errors were chosen to be the nominal Phase I values). Best results were obtained using a scale factor of .001, and this was used in the last set of Figures M through O. In this latter set, again R_k was manually decreased for 3 of the 6 ot measurement variables. Overall, the best results (smallest estimate errors) are shown in Figures M and O, for the case $R_k(j,j) = 10^{-1}$ for ot variables. Adaptation by direct feedback of $P_k(+)$ is a less sophisticated means of estimating Qk in real time than the method of maximum likelihood albeit, it is much simpler to implement.

3.0 Estimates of Technical Feasibility

VLSI (Very Large Scale Integration) technology has been developed to the point where high speed floating point processors may be concatenated to form

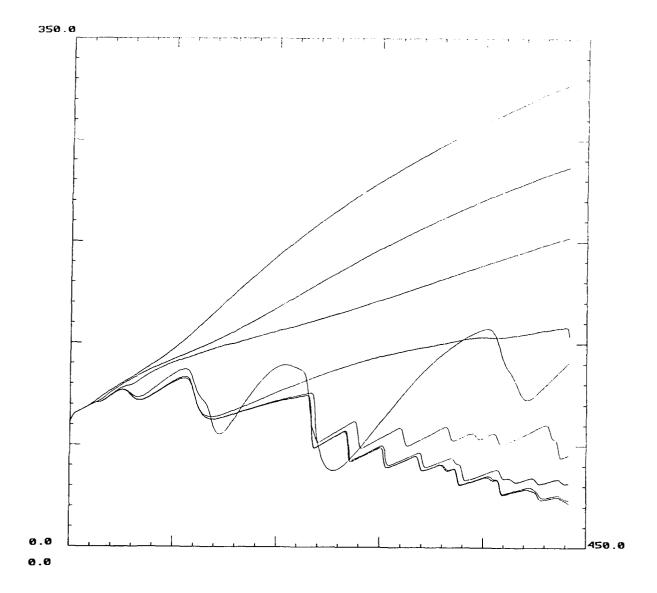


Figure A

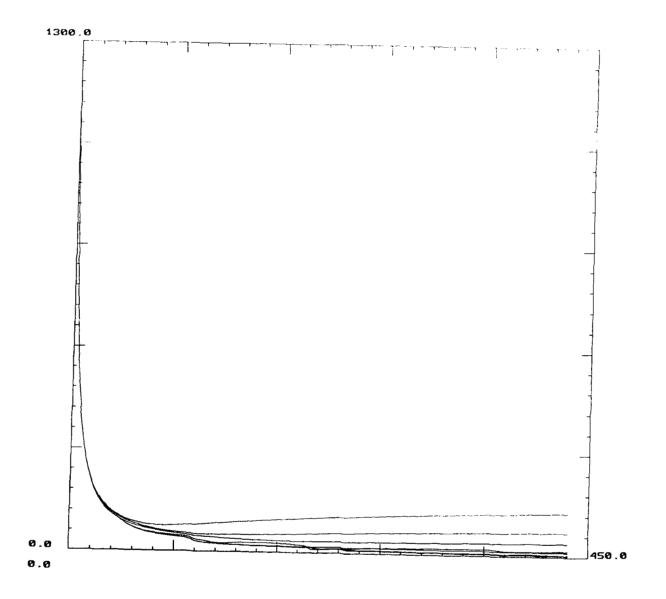


Figure B

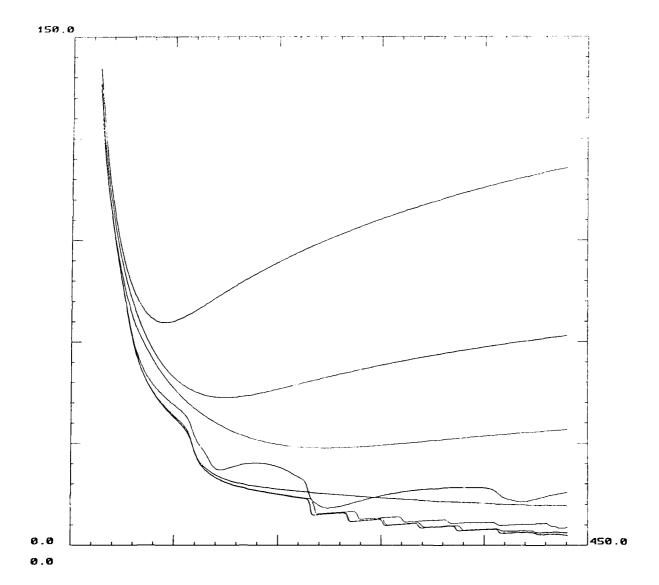


Figure C

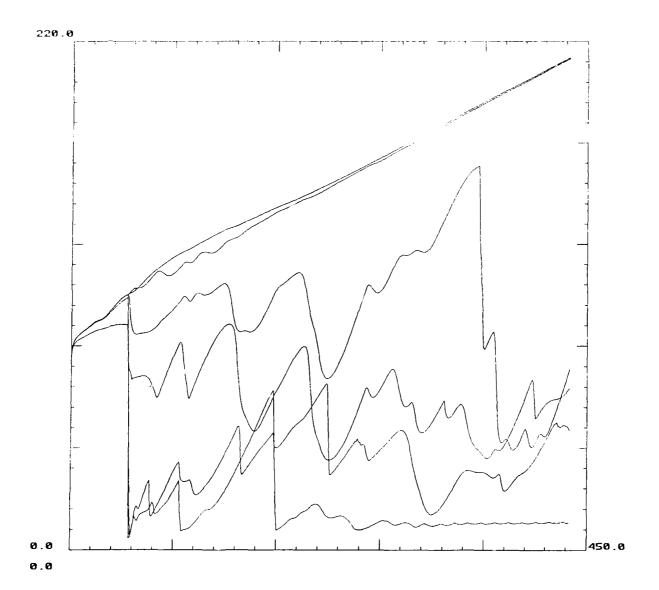


Figure D

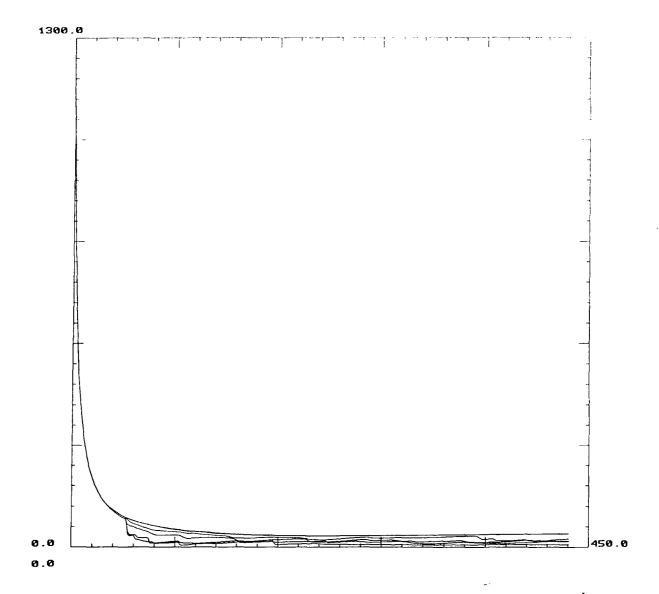


Figure E

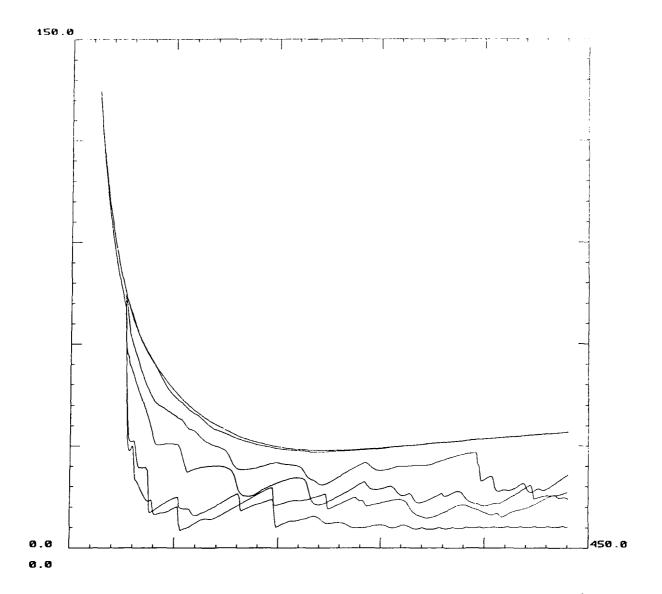


Figure F

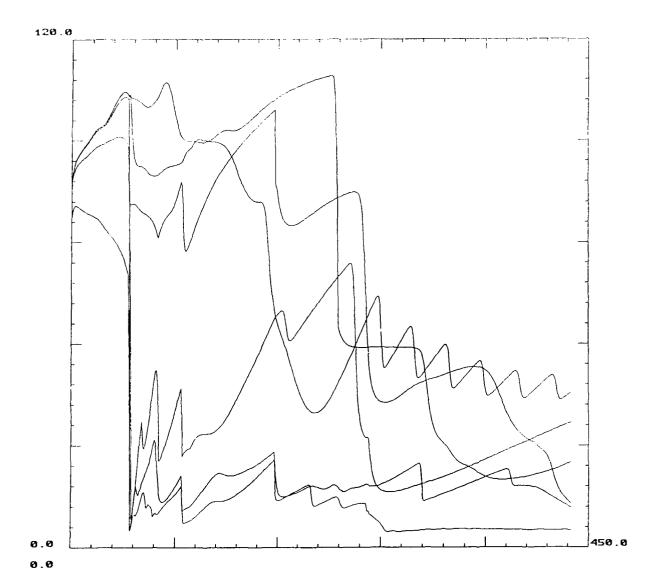


Figure G

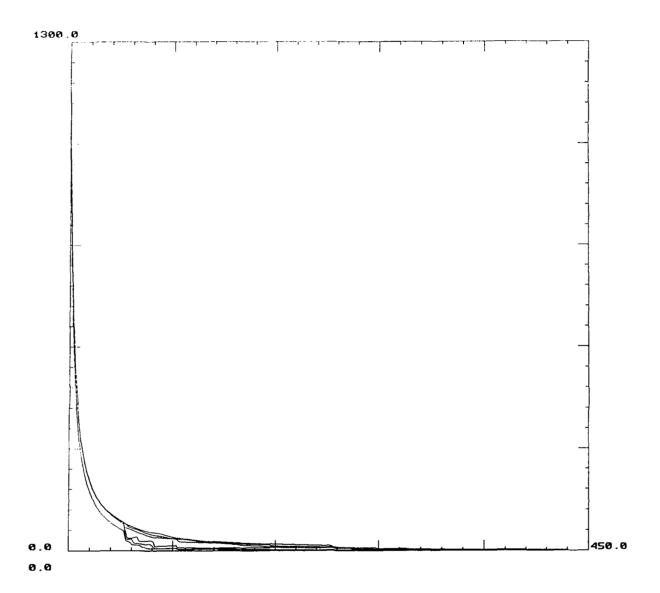


Figure H

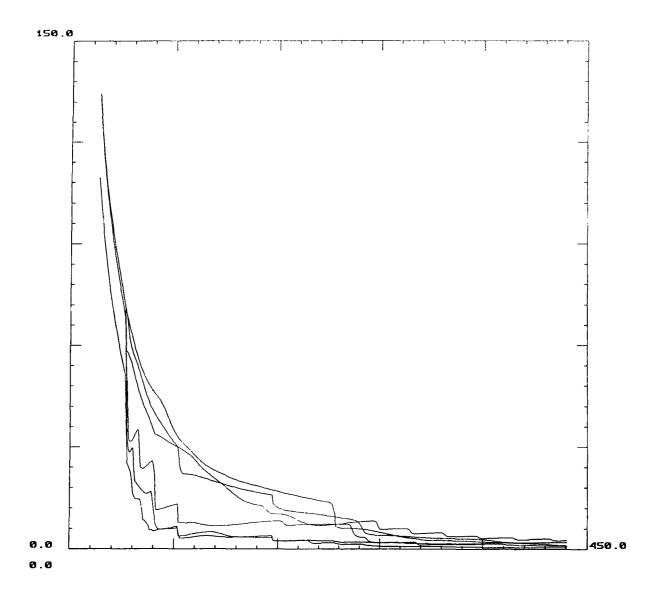


Figure I

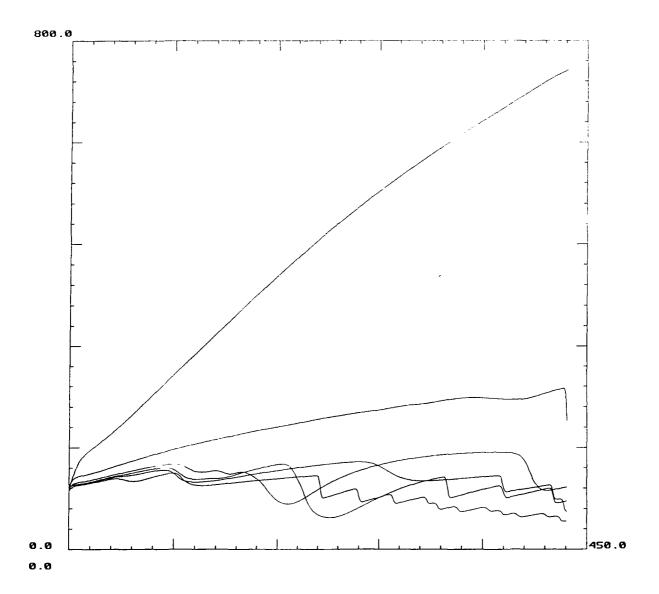


Figure J

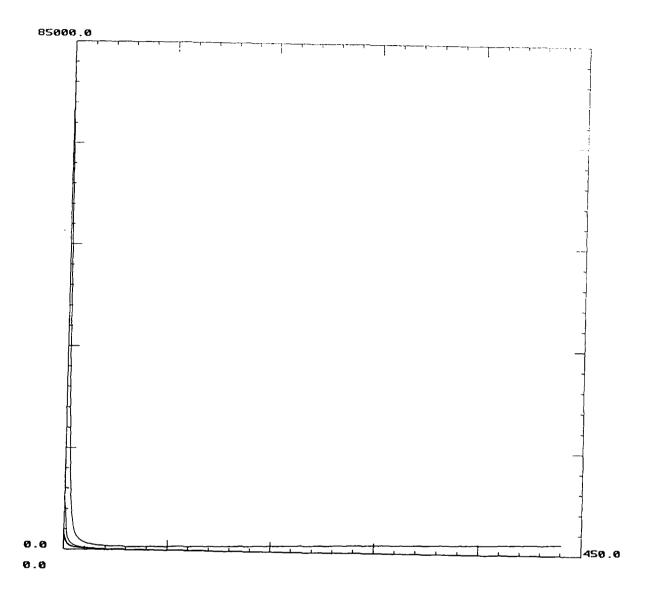


Figure K

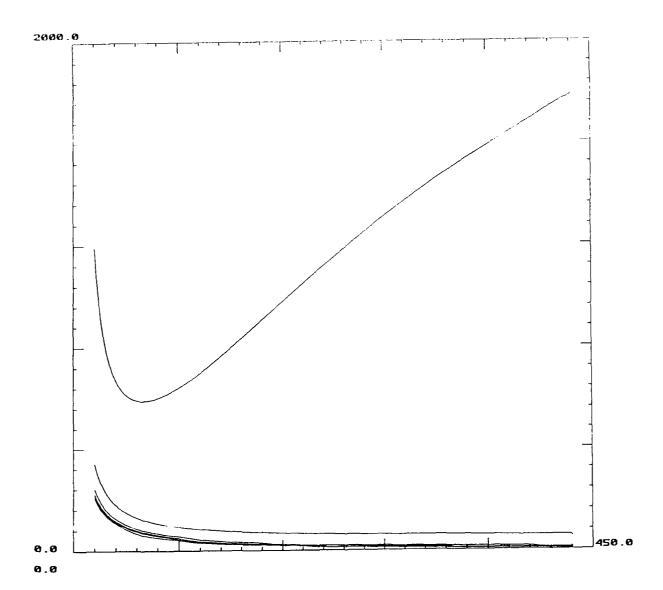


Figure L

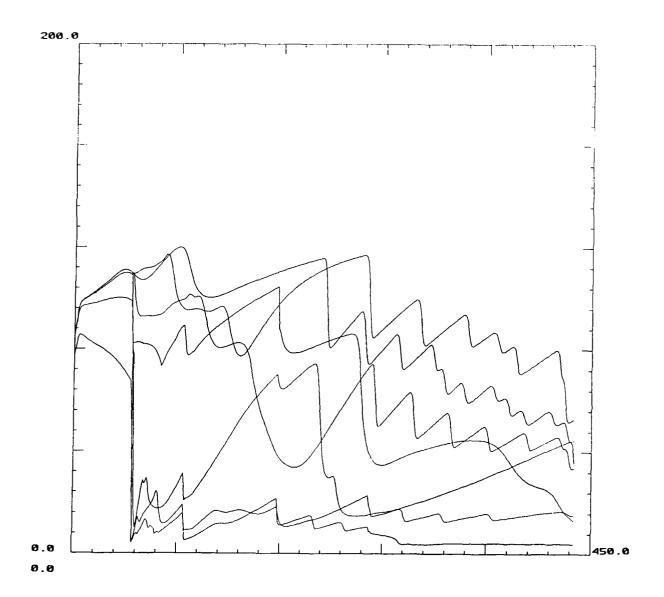


Figure M

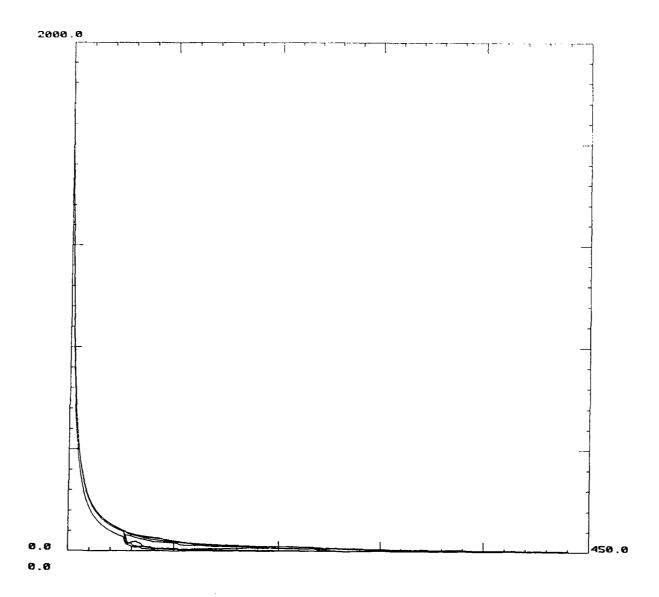


Figure N

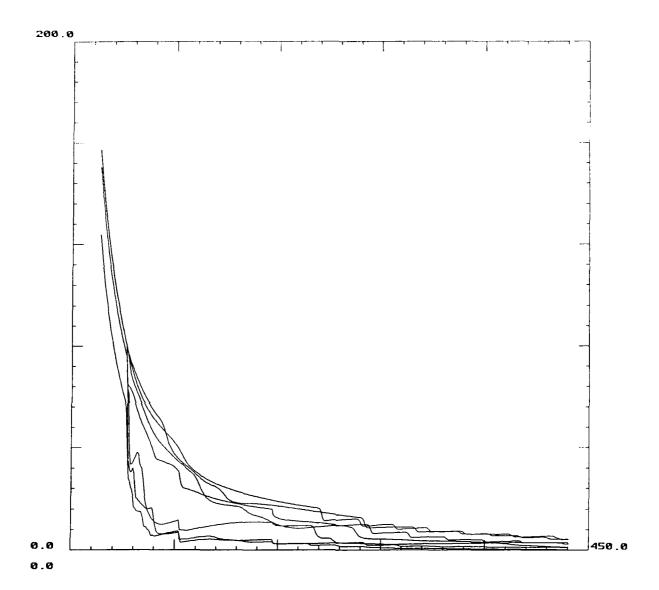


Figure 0

compact supercomputers with far greater throughput than uniprocessor machines. Thus, there is considerable interest within the signal processing community in the development of parallel versions of conventional algorithms. MTI has collaborated with Dr. G.J. Bierman to develop a parallel form of the Kalman filter that has several very unique and important features. We believe that utilization of these features will result in the design of an integrated tracking system that exhibits much improved performance over any "isolated" approach to undersea surveillance or test range tracking.

Specifically, our Decentralized Square Root Information Filter (DSRIF) [1] allows each group of measurement variables, the process noise statistics and the prior information about the initial state to be processed in separate but locally optimal filters. Globally optimal state estimates and estimate error covariances may then be computed by combining local filter outputs on demand. This will allow the analyst to identify the contribution of each measurement group, the process noise and the prior information about the initial state to the global state estimate and estimate error covariance without additional computation.

Furthermore, the process noise and prior information may be distributed amongst the data processing filters in order to improve upon the fault tolerant characteristics of the nominal algorithm when real-time signal processing is an issue. In this case, the estimates and covariances should gracefully degrade from global optimality as local processors fail. Thirdly, the algorithm is based upon numerically reliable matrix factorization methods which, unlike the CKF, will never fail.

The objective of Phase I research was to validate the DSRIF equations by testing its ability to track both predetermined and unknown trajectories when perturbed by white Gaussian noise. The state estimates and error covariances obtained were found to be identical (when printed to 10 significant digits) with those of a SRIF implemented in centralized form with all calculations performed in double precision arithmetic. Furthermore, an adaptive version of the extended DSRIF (E-DSRIF) was successfully used to track real Multiple Rocket Launch System data obtained from the WSMR.

In order to determine the feasibility of our distributed approach to multisensor tracking, several specific technical objectives must be met. First and foremost, the basic DSRIF theory needs to be extended to enable the tracking of multiple targets. This requires that a theory for associating data with targets, based upon the DSRIF (and not the SRIF which already exists, see [2]), be developed. Correlation of measurements with targets can best be done using a hypothesis testing approach. The idea is to select the correlation of measurements with targets that has maximum probability given the data. Calculation of all combinations to form the entire set of these conditional probabilities can be prohibiting, especially in a dense target environment. A major advantage in using the DSRIF is the tremendous reduction in computational cost associated with this calculation.

Secondly, the DSRIF is a new algorithm which has undergone only limited testing in Phase I research. Extensive testing within a multisensor multitarget tracking scenario is needed. Other theoretical questions such as the development of a delayed-state DSRIF for processing range-rate measurements, a method for isolating faulty sensors, and efficient

implementations of the DSRIF that facilitate high data rates need to be addressed.

Finally, consideration needs to be given to the design of the tracking network both at the global and local levels. The major question here is whether a sufficient data rate can be achieved using current chip technology. Another question is whether the architecture can be reconfigured (in software) to implement other members of the family of DSRIFs. A multitude of test range scenarios is envisioned so that a robust system is needed. At one extreme, test vehicles may include slow moving submarines with well defined nominal trajectories a priori while at the other, multiple smart torpedoes with maneuvering capability is possible. The key to a successful network design is to employ a more or less sophisticated version of the algorithm depending upon the particular scenario. Thus the network must be adaptable. For example, preflight simulations of the proposed shot using high fidelity hydrodynamic models can yield good values for the process noise levels and a basic DSRIF should result in good tracking performance. However, a sudden departure from the nominal trajectory would require a detection mechanism as part of the algorithm and adjustment of $\mathbf{Q}_{\mathbf{k}}$ in real time. A DSRIF based multisensor laboratory tracking experiment should be performed prior to deployment of a test range prototype.

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